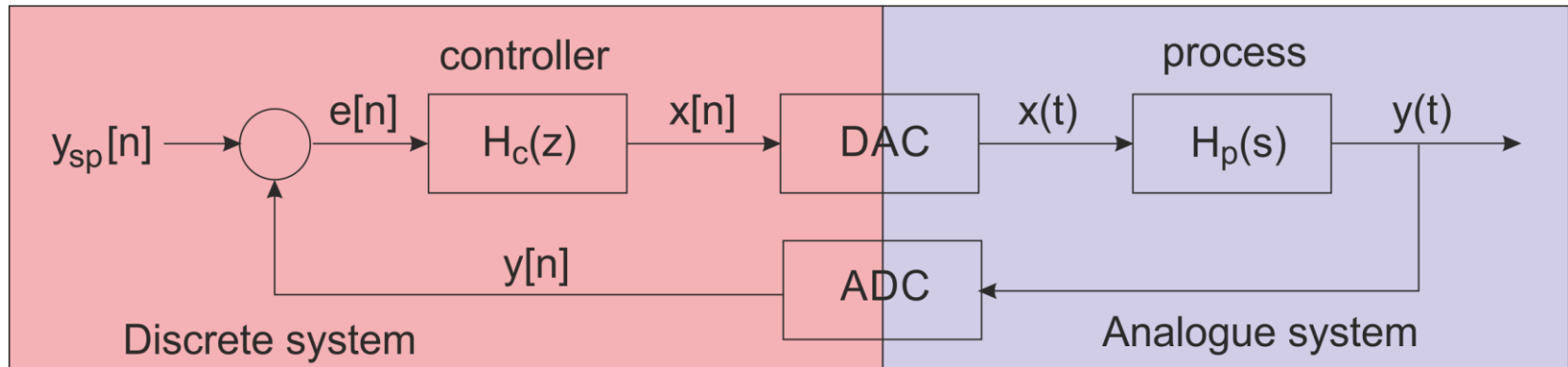


Lesson 5

- **Combination of analogue and digital systems**
- **Relation between s (Laplace) and z (discrete)**
- **Euler approximation**
- **Tustin approximation**
- **Questions / exercises**

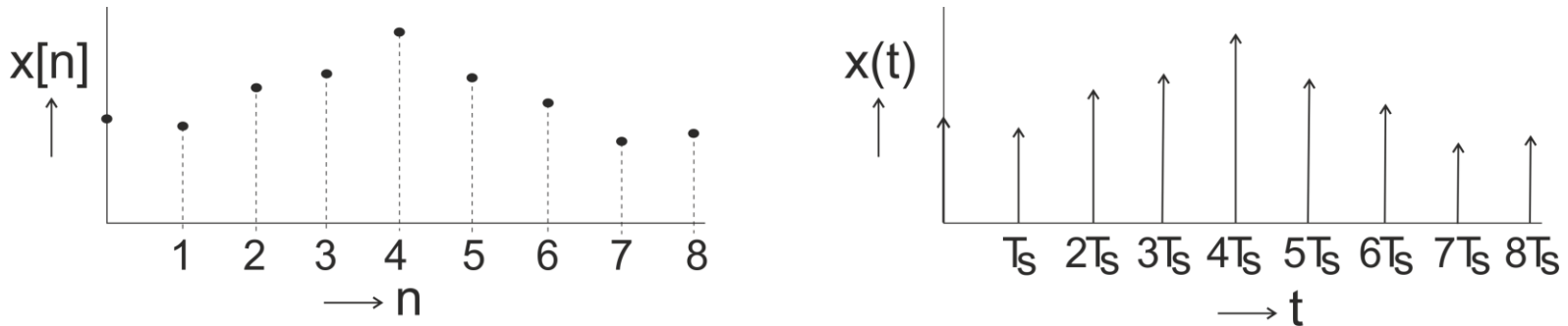
Combination of discrete and analogue systems



$$Y(z, s) = \frac{H_c(z) \cdot H_p(s)}{1 + H_c(z) \cdot H_p(s)} \cdot Y_{sp}[z] \quad ???$$

Relation between z and s

Consider a function $x(t)$ that is sampled with a sample time of T_s .



The z-transform of $x[n]$ can be written as: $X(z) = \sum_{n=0}^{\infty} x[n] \cdot z^{-n}$

The discretized signal $x(t)$ can be written as a sum of dirac functions $\delta(t)$:

$$x(t) = \sum_{n=0}^{\infty} x(n \cdot T_s) \cdot \delta(t - n \cdot T_s) \rightarrow X(s) = \int_0^{\infty} \sum_{n=0}^{\infty} x(n \cdot T_s) \cdot \delta(t - n \cdot T_s) \cdot e^{-s \cdot t} \cdot dt \rightarrow$$
$$X(s) = \sum_{n=0}^{\infty} x(n \cdot T_s) \cdot e^{-s \cdot n \cdot T_s} = \sum_{n=0}^{\infty} x[n] \cdot e^{-s \cdot T_s \cdot n}$$



X(z) and X(s) equal if: $z = e^{s \cdot T_s} \rightarrow s = \frac{\ln(z)}{T_s} = f_s \cdot \ln(z) \quad f_s = \frac{1}{T_s}$

Analogue systems can be combined with discrete systems by the substitution of s by: $s \rightarrow f_s \cdot \ln(z)$.

However, this substitution has a big disadvantage:

the numerator and denominator of the overall transfer function are no polynomial functions anymore. This makes a the system description in terms of poles and zeros, stability, bode diagrams, etc. as you learned it impossible.

Euler gives an approximation of $z = e^{s \cdot T_s}$ in such a way that the overall transfer function remains a quotient of 2 polynomials, so all theorie discussed remains valid.

Forward Euler (1)

Assume the function: $s = f_s \cdot \ln(z)$

The first order Taylor-series approximation of s around $z=1$ equals:

$$s \approx s(z = 1) + \left. \frac{ds}{dz} \right|_{z=1} \cdot (z - 1) = f_s \cdot (z - 1)$$

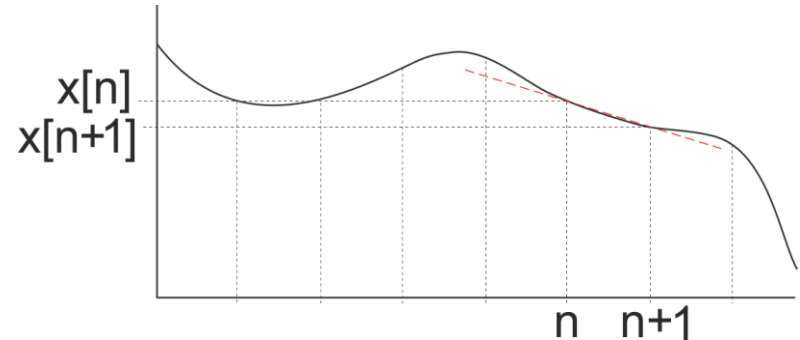
This approximation is called the forward difference Euler approximation (or simply forward Euler).

Forward Euler (2)

Another way to look at it:

A derivation in the time domain corresponds to a multiplication of s in the s -domain. So:

$$y(t) = \frac{dx}{dt} \rightarrow Y(s) = X(s) \cdot s$$



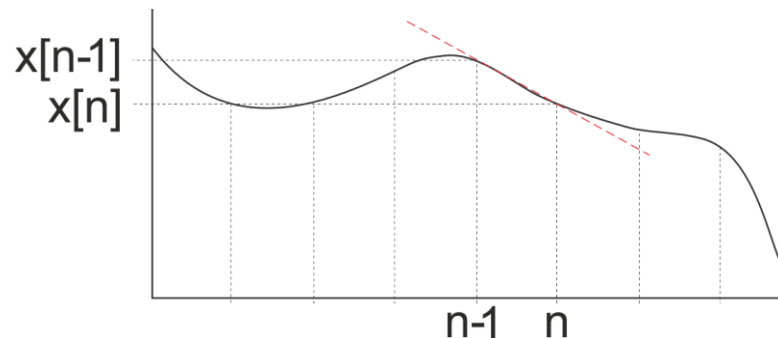
In the time-domain, we can approximate a derivation by:

$$y(t) = \frac{dx}{dt} \approx \frac{x[n+1] - x[n]}{T_s} \rightarrow Y(z) = \frac{z \cdot X(z) - X(z)}{T_s} = X(z) \cdot f_s \cdot (z - 1)$$

We see that a derivation in the time domain corresponds to a multiplication with $f_s \cdot (z-1)$ in the z -domain. This means that the Laplace operator s corresponds to $f_s \cdot (z-1)$ in the z -domain

Backward Difference Euler

Instead of looking one sample ahead, you can also approximate the derivation of $x(n \cdot T_s)$ by looking one sample back:



$$y(t) = \frac{dx}{dt} = \frac{x[n] - x[n-1]}{T_s} \rightarrow Y(z) = \frac{X(z) - \frac{X(z)}{z}}{T_s} = X(z) \cdot f_s \cdot (1 - z^{-1})$$

We see that a derivation in the time domain corresponds to a multiplication with $f_s \cdot (1 - z^{-1})$ in the z-domain. This means that the Laplace operator s corresponds to $f_s \cdot (1 - z^{-1})$ in the z-domain

Example Euler

Assume a second order Butterworth low pass filter, given by:

$$H(s) = \frac{1}{\left(\frac{s}{\omega_0}\right)^2 + \sqrt{2} \cdot \frac{s}{\omega_0} + 1}$$

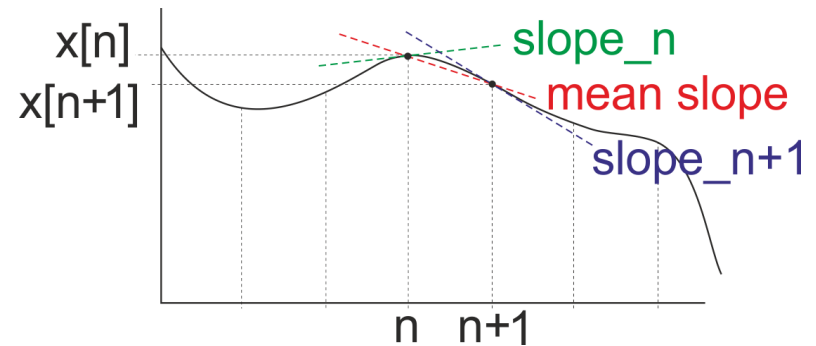
The corresponding discrete transfer function $H(z)$ of it is given by:

$$H(z) = \frac{1}{\left(\frac{f_s}{\omega_0}\right)^2 \cdot (1 - z^{-1})^2 + \sqrt{2} \cdot \frac{f_s}{\omega_0} (1 - z^{-1}) + 1} \rightarrow$$

$$H(z) = \frac{z^2}{\left(\left(\frac{f_s}{\omega_0}\right)^2 + \sqrt{2} \cdot \frac{f_s}{\omega_0} + 1\right) \cdot z^2 - \left(2 \cdot \left(\frac{f_s}{\omega_0}\right)^2 + \sqrt{2} \cdot \frac{f_s}{\omega_0}\right) \cdot z + \left(\frac{f_s}{\omega_0}\right)^2}$$

This equation can be used to derive the difference equation.

Tustin provides a more accurate approximation of s . It assumes that the quotient $\frac{x[n+1]-x[n]}{T_s}$ is the average of the derivative of $x(t)$ at $t=n \cdot T_s$ and $t=(n+1) \cdot T_s$.



$$y(t) = \frac{dx}{dt} \quad \text{and:} \quad \frac{x[n+1] - x[n]}{T_s} = \frac{y[n+1] + y[n]}{2} \quad \rightarrow$$

$$z \cdot Y(z) + Y(z) = \frac{2}{T_s} (z \cdot X(z) - X(z)) \quad \rightarrow \quad Y(z) = X(z) \cdot 2 \cdot f_s \cdot \frac{z-1}{z+1}$$

A derivation in the time domain corresponds to a multiplication with $2 \cdot f_s \cdot (z-1)/(z+1)$ in the z -domain. According to Tustin, the Laplace operator s corresponds to $2 \cdot f_s \cdot (z-1)/(z+1)$ in the z -domain.

Assume a first order high pass filter, given by:

$$H(s) = \frac{\frac{s}{\omega_0}}{\frac{s}{\omega_0} + 1}$$

The corresponding discrete transfer function $H(z)$ using Tustin is given by:

$$H(z) = \frac{2 \cdot \frac{f_s}{\omega_0} \cdot \frac{z-1}{z+1}}{2 \cdot \frac{f_s}{\omega_0} \cdot \frac{z-1}{z+1} + 1} = \frac{2 \cdot \frac{f_s}{\omega_0} \cdot (z-1)}{\left(2 \cdot \frac{f_s}{\omega_0} + 1\right) \cdot z + 1 - 2 \cdot \frac{f_s}{\omega_0}}$$

This equation can be used to derive the difference equation.

Questions / Exercises

1. Give the transfer function $H(z)$ and difference equation of the following analogue systems. Use Euler backward:

a. $H(s) = \frac{1}{1+s\cdot\tau}$ where: $\tau = 0.01$ [s]. The sample rate is: $f_s=1$ [kHz]

b. $H(s) = \frac{1}{1+s\cdot\tau}$ where: $\tau = 0.01$ [s]. The sample rate is: $f_s=100$ [kHz]

c. $H(s) = \frac{s\cdot 0.1}{(1+s\cdot 0.02)\cdot(1+s\cdot 0.005)}$. The sample rate is: $f_s=100$ [kHz]

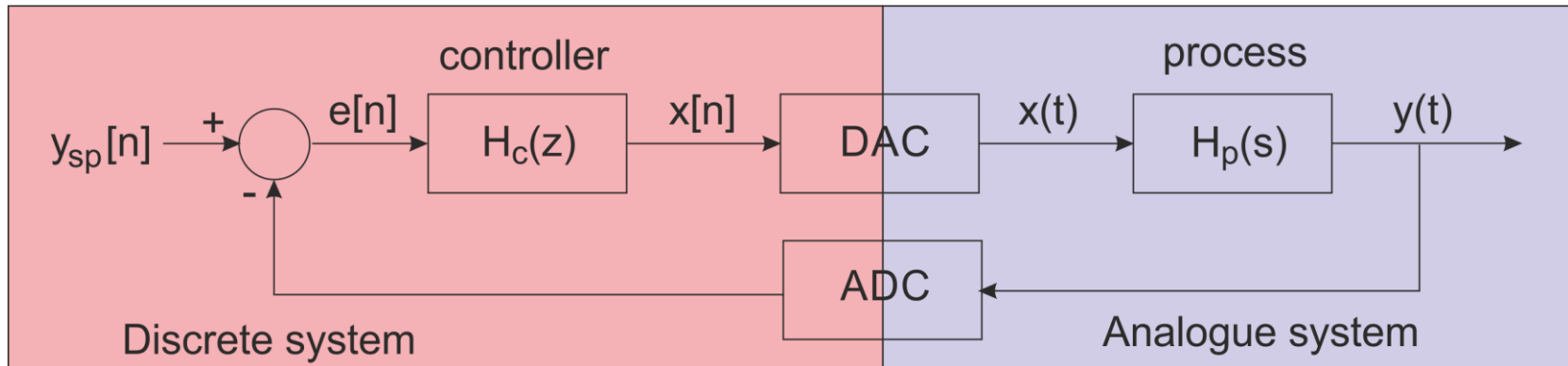
2. Given a PID controller: $H(s) = 5 + 2 \cdot s + \frac{0.3}{s}$

Give the transfer function $H(z)$ and the difference equation at a sample rate of 500 [samples/s]. Use Tustin.

Questions / Exercises

3. A first order (analogue) process is described by: $H_p(s) = \frac{6}{1+s \cdot 0.5}$

The process is controlled via a discrete P-controller (see figure) with a proportional constant of 5. The sample rate is 20 [samples/s].



a. Give the overall transfer function of the system, defined as:

$H(z) = Y(z) / Y_{sp}(z)$. Use Tustin.

b. Calculate the step response $y[n]$ ($y_{sp}[n] = u[n]$).

Questions / Exercises

4. Consider the system of question 3, only now we control the process via a discrete PI-controller. The proportional constant is 5, the sample rate is 20 [samples/s]. The integral constant is chosen in such a way that the system is critically damped.
- Describe when we can speak of a critically damped system.
 - Calculate the gain of the I-controller.
 - Give the difference equation of the discrete PI-controller.
 - Show in a graph the discrete step response $y[n]$.